



AM Electronic Tuner

Overview

LA1245 is a high performance IC to be used as an AM electronic tuner. It provides an automatic search-stop signal, local oscillator buffer-output, and the low level local oscillation, as well as providing all other functions required of an AM tuner. Moreover, the stable local oscillation from LW to SW facilitates the use of many band.

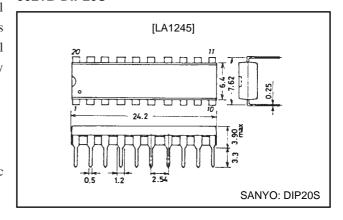
Functions

- RF amplifier MIX OSC (with ALC) Detection
- IF amplifier AGC Local oscillation buffer-output
- Signal meter driving output (also used as an automatic search stop-signal)
- etc.

Package Dimensions

unit: mm

3021B-DIP20S



Features

• Narrow-band signal meter : Available as an automatic search-stop signal (also available as a wide-band signal

meter). Signal meter output=1/2 frequency ± 1.5 kHz typ.

• Local oscillation buffer-output : Facilitates the design of electronic tuning systems and frequency representation.

• OSC (with ALC) : The oscillation output is stabilized at a low level (350 mVrms) for a varactor

diode, and tracking error is minimized.

• RF amplifier : Excellent in usable sensitivity by incorporating low-noise transistors in cascode

circuit (45dB/m typ).

• MIX : Double balanced differential MIX prevents the influence of spurious radiation and

IF interferences (IF interference = 85dB typ).

• Low noise : Excellent in S/N for intermediate input (57dB typ).

 \bullet Compensation for V_{CC} fluctuation : Allows little gain fluctuation and little distoriton fluctuation (8 to 16 V).

• Low shock noise : Able to decrease the shock noise by selecting AGC time constant when changing

V_{CC}-on and/or switching the mode.

Specifications

Maximum Ratings at Ta=25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V _{CC} max	Pin 8, 14	16	V
Output voltage	Vo	Pin 5, 7	24	V
Input voltage	VI	Pin 3	5.6	V
Supply current	I _{CC} max	Pin 5+7+8+14	32	mA
Output high drive current	I ₁₈	Pin 18	5	mA
	I ₂₀	Pin 20	2	mA
Allowable power dissipation	Pd max	See Figure 2	700	mA
Operating temperature	Topr		-20 to +70	°C
Storage temperature	Tstg		-40 to +125	°C

Recommended Operating Conditions at Ta=25°C

Parameter	Symbol	Conditions	Ratings	Unit
Recommended supply voltage	V_{CC}		12	V

Operating Characteristics at Ta=25°C, V_{CC} =12V, f_r =1MHz, f_m =400Hz, at specified test circuit (based on application circuit).

Parameter	Symbol	Conditions	Ratings			Unit
1 drameter	Symbol		min	typ	max	Oille
Current drain	I _{CC} 1	quiescent	16.0	25.0	35.0	mA
	I _{CC} 2	107 dBµ input	19.0	29.0	40.0	mA
Detection output	V _O 1	23 dBµ input, mod. 30%	-27.5	-23.0	-18.5	dBm
	V _o 2	80 dBµ input, mod. 30%	-15.5	-12.5	-9.5	dBm
Signal to noise ratio	S/N1	23 dBµ input, mod. 30%	16	20		dB
	S/N2	80 dBµ input, mod. 30%	52	57		dB
Total harmonic distortion	THD1	80 dBµ input, mod. 30%		0.4	1.0	%
	THD2	107 dBµ input, mod. 30%		0.3	1.0	%
Signal meter output	V _{SM} 1	quiescent		0	0.5	V
	V _{SM} 3	107 dBµ input	3.0	4.5	7.0	V
Input at signal meter output=1V	V _{IN} 1	V _{SM} output=1V	19.0	25.0	31.0	dΒμ
Local oscillation-buffer output	Vosc		250	350		mVrms

Reference Characteristics

Parameter	Symbol	Conditions	Ratings			Unit
i alametei	Symbol		min	typ	max	Offic
Signal meter output	V _{SM} 2	40 dBμ input		2.5		V
Total harmonic distortion	THD3	112 dBµ input, mod.30%		2		%
Local oscillation fluctuation	ΔV_{OSC}	V _{osc} L (522kHz) to V _{osc} H		10		mVrms
within a band		(1647kHz)				
Signal meter band width*	V _{SM-BW1}	80 dBµ input, 1/2 output frequency		±1.5		kHz
	V _{SM-BW2}	80 dBµ input, 1/10 output frequency		-4.5/+7		kHz
Selectivity		±10kHz at 30% mod.		45		dB
IF interference		f _r =600kHz		85		dB
Image frequency interference		f _r =1400kHz		40		dB
ratio						

^{*} BFB450C4 N (Murata, Co.,) was used as a narrow band filter. (Note) 0 dBm=775mV, 0 dBu=1 $\mu V.$

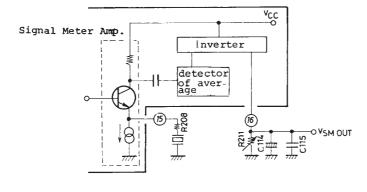
Using the automatic search-stop signal

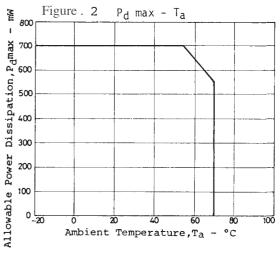
Signal Meter-driving output circuit is equivalent to Figure. 1, signal meter driving output (abbreviated as V_{SM}) is narrowed in band width and can be used as an automatic search-stop signal when a narrow band series resonator is connected to pin 15. V_{SM} can be adjusted with R_{208} and R_{211} both in wide band and narrow band since R_{208} is inversely proportional to V_{SM} , while R_{211} is proportional to V_{SM} . R_{208} is related to the Q of narrow band signal meter. When the resistance of R_{208} is increased, the Q will be damped and the band width increased. On the other hand, R_{211} used as the output impeadance of V_{SM} and affects the cut-off frequency and time constant of low pass filter for V_{SM} and the meter drive impedance. The time constant τ and the cut-off frequency fc can be expressed as follows:

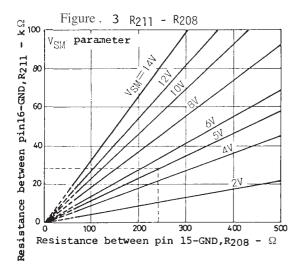
$$\left\{ \begin{array}{l} \tau = (C_{114} + C_{115} + C_S) \ (R_{211} / / R_{in}) \\ f_c = \frac{1}{2\pi\tau} \end{array} \right.$$

A semi-fixed resistor is recommended to be used as R_{211} to cope with the fluctuation of V_{SM} . Refer to Figure. 3 for the value of the semi-fixed resistor since this depends upon V_{SM} and R_{208} . Figure. 3 shows the lowest limit of the semi-fixed resistor in relation to R_{208} with the parameter of V_{SM} set point, and the value of the semi-fixed resistor will be equal to or greater than that shown in Figure. 3. For example, when V_{SM} =5V and R_{208} =240 Ω , R_{211} becomes $28k\Omega$. Thus, the value of the semi-fixed resistor is determined to be about $30k\Omega$. When the value of V_{SM} is too large, it is limited and saturated to the source voltage so it is recommended to follow the condition of $V_{SM} \le V_{CC}$ -2(V). When a narrow band serial resonator is used, include the resonant impedance to determine the value of R_{208} .

Figure . 1 Signal Meter Detector Circuit

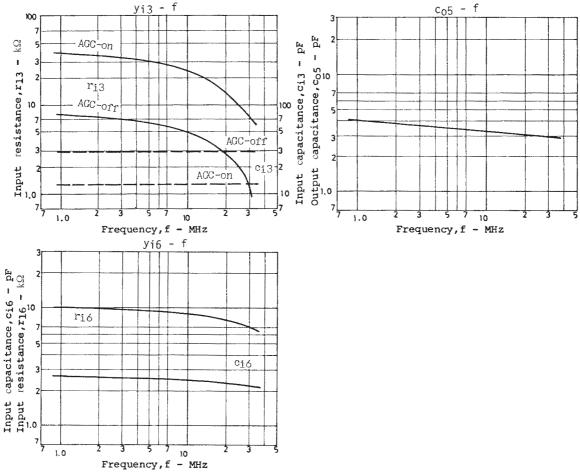






Input/Output Admittance

	Parameter	Frequency	_	AGC-off	AGC-on
RF	УіЗ	1 MHz	r _i c _i	8 kΩ 30 pF	40 kΩ 13 pF
T.T.	У05	1 MHz	r _o c _o	— 4 pF	_
MIX	Уіб	1 MHz	r _i c _i	10 kΩ 2.6 pF	_
MITV	У07	500 kHz	r _o c _o	— kΩ 2 pF	_
1st IF	Уi9	500 kHz	r _i c _i	3 kΩ 7 pF	3.2 kΩ 3 pF
	У010	500 kHz	r ₀ c ₀	45 Ω 20 pF	42 Ω 20 pF
2nd IF	УіП	500 kHz	r _i c _i	80 Ω -150 pF	_



Notes on LA1245 usage

1. When suddenly tuned to a broadcasting station of intermediate or high field strength, a large current of high frequency flows into the signal meter circuit, causing the local oscillator malfunctions and abnormal noises.

To eliminate this:

- · Use $R_{208} \ge 240\Omega$ for manual tuning type.
- · Use $R_{208} \ge 82\Omega$, and use the local oscillation coil at the 1/3 tap (except SW) for electronic tuning type (which uses a narrow band filter).
- 2. Use the bias on the condition RF $V_{CC} \le IF V_{CC}$, since abnormal noise levels might be caused when detuning a strong input on the codition RF $V_{CC} > IF V_{CC}$.
- 3. Use the signal meter driving output (V_{SM}) at $V_{SM} \le V_{CC} 2$ (V) to avoid saturation caused by V_{CC} .
- 4. Use 1/2 or more tap of LW and MW oscillation coil to improve S/N and the detuning characteristics of the distortion ratio.
- 5. Use the full-tap of SW oscillation coil, to allow the sag in oscillation power by the decreasing of Q.
- 6. Avoid the coupling of the antenna tuning circuit and the local oscillating circuit so as not to leak the local oscillation into the antenna tuning circuit.
- 7. Connect the detection capacitor C_{113} between pin 13 (output) and pin 14 (V_{CC}) to avoid the leakage of the IF signal into the GND line. Connection between pin 13 and pin 12 (GND) increases the tweet interference and deteriorates the usable sensitivity.

Moreover, depending on the positions of C_{113} and the bar antenna, higher harmonics having twice or three times the frequency of the IF signal may pass into the antenna and cause tweet interference, and in extreme cases oscillation might be cause. To prevent this:

- · Shorter lead wires and connect them near 13 and 14 pins.
- \cdot Place C_{113} far from the antenna.

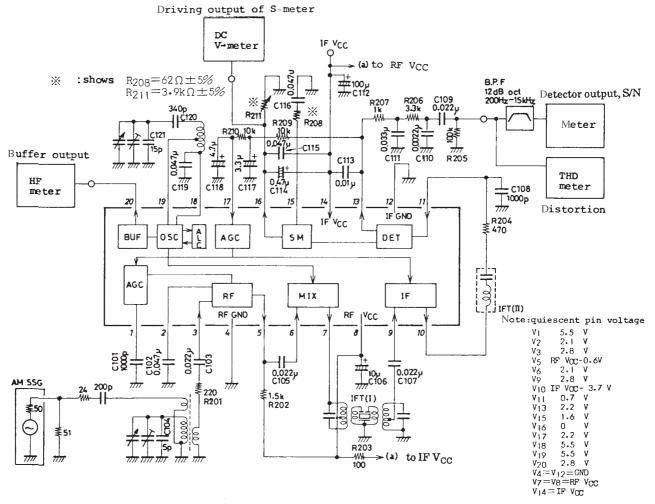
- 8. When a cable or something similar is connected to a local oscillation buffer (pin 20), which is equivalent to connecting a capacitor of about 20pF, the output from the buffer will be of sawtooth waves, causing the level low at the short wave band. To prevent this, connect a resistor between pin 20 and GND, which will increase the operating current of the buffer amplifier. Since the maximum current obtained from pin 20 is 2mA, the suitable resistance between pin 20 and GND is 1.5kΩ.
- 9. Use a semi-fixed resistor for R_{211} to allow the fluctuation of V_{SM} .
- 10. When changing an IFT or using an RF tuner, select a filter and related circuits according to the following conditions. The input levels of each terminal where 30% modulated detection output of –25dBm is obtained are as follows:

Pin 11 input	when Rg= 520Ω ($470\Omega + 50\Omega$)	75dBµ
Pin 9 input	when Rg= 50Ω	53dBµ
Pin 6 input	when Rg= 50Ω	48dBμ
Pin 3 input	when Rg= 50Ω	22dBµ

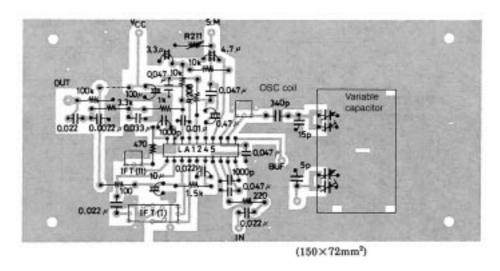
Slight change in IFT, however, will be covered by changing the constant of resistors R₂₀₂ and R₂₀₄.

11. When the coupling coefficient of the local oscillation coil is small and an anti-resonance point of about 100MHz is present or the stray capacitance between pin 19 and pin 20 is large, the buffer output (pin 20) may be subject to parasitic oscillation of about 100MHz. In this case, connect a capacitor of about 30pF between pin 20 and GND. To observe parasitic oscillation, connect a capacitor of 5pF in series with the probe. If the probe is connected direct to pin 20, the input capacitance of the probe causes parasitic oscillation to stop, which makes it impossible to observe.

Sample Application Circuit 1

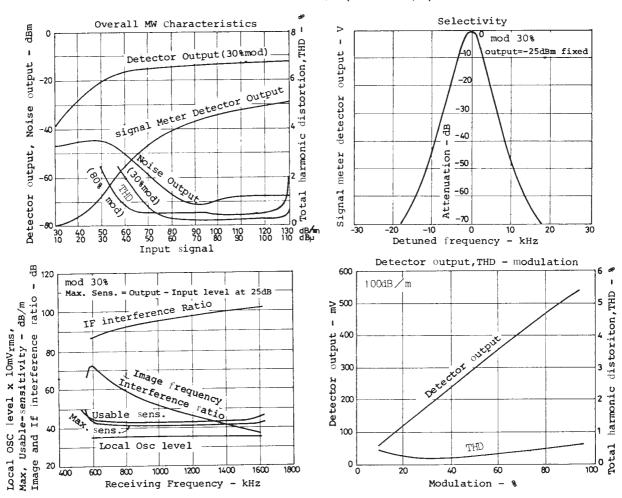


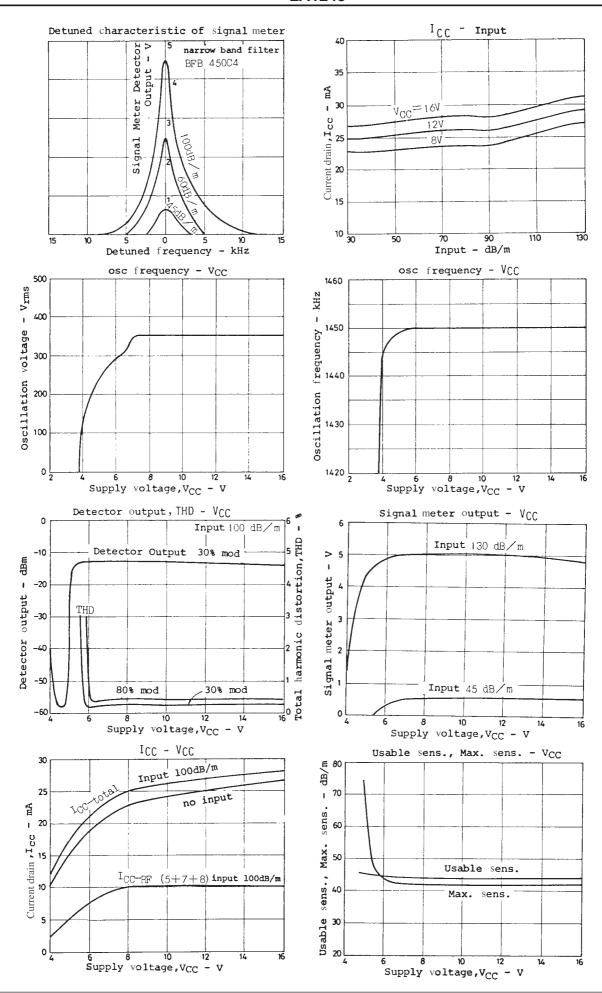
Unit (resistance : Ω , capacitance : F)

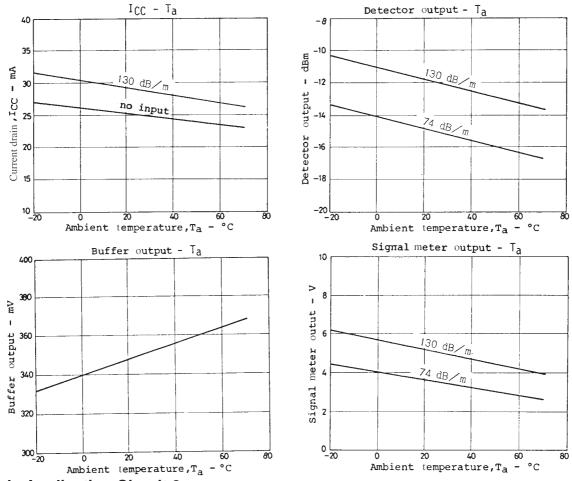


An Example of Printed Pattern (150x72mm², bottom view)

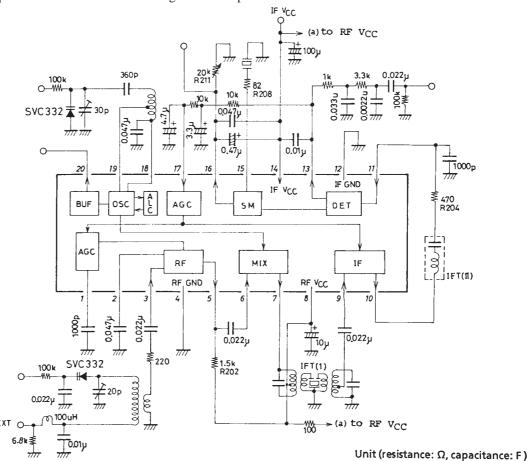
Unit (resistance: Ω, capacitance: F)

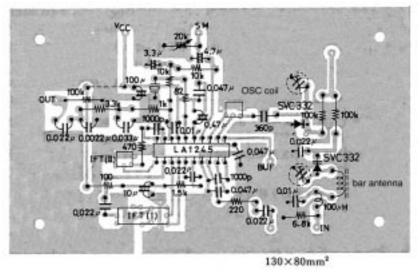


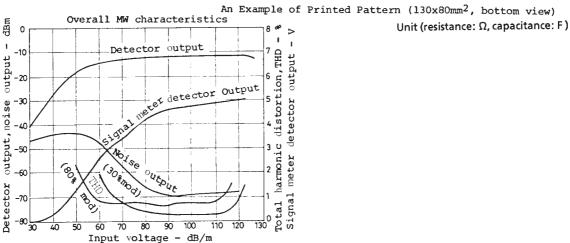




Sample Application Circuit 2: Using variable capacitance diodes

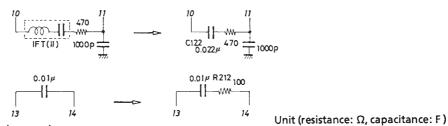






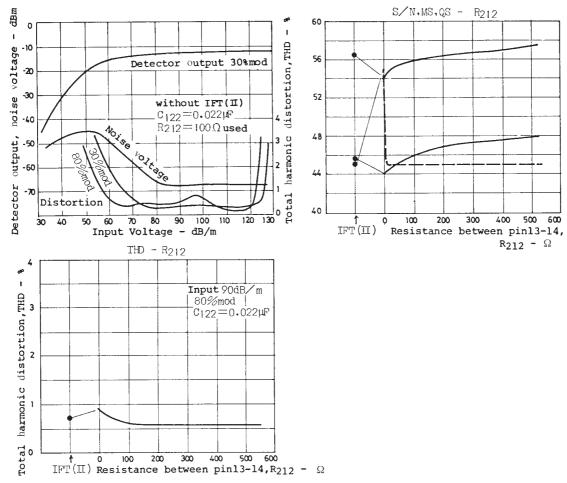
Sample Application Circuit 3: Rejecting IFT (II)

Following 2 changes are recommended as C-conpling without IFT (II)



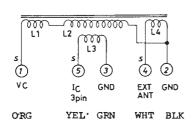
Comparison of characteristics varying parts.

Using IFT(II) C122=0.022µF dBm dBm 0 t 0 distoriton, THD harmonic distortion, THD -10 - 10 output, noise voltage voltage Detector output 30% mod Detector output 30%mod -20 45.5 dB/m 45.0 dB/m noise Input 90 dB/m Without IFT(II) S/N 56.5 dB C-conpled harmonic output, 1015a36 - 60 Detector Detector Total Distortion Distortion 60 70 80 90 100 110 Input voltage - dB/m 50 60 80 90 100 110 120 130 Input voltage - dB/m



Peripheral Parts

(1) Bar Antenna (34H-052-869 Sumida Co.,)



For use of general variable capacitor

L(between pins 1,2)=270µH

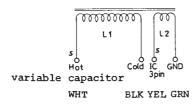
Q≥180

L|: solenoid 43 t. L2: space 42 t.

L3: solenoid 7 t.

L4: solenoid

(2) Bar Antenna (C-4698 Coil Snake Co.,)



For use of variable capacitor diode

L(between pins 1,2)=250µH

Q≥250

L₁: solenoid 55 t.

L2: solenoid 5 t.

(3) Osc coil



2157-223-072 Sumida

2157-223-082 Sumida 7BR-6654Y Toko

L(between pins 1 and 3)=147µH L(between pins 1 and 3) 147µH L(between pins 1 and 3)=147µH Q≧90 ③② ②① Q≥85 **3**•2 31 t.

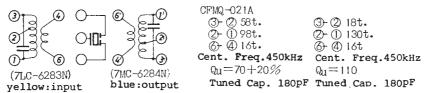
31 t.

(4) Variable Capacitor (C123A Alps Co.,)

c max 326.8 pF c min 6.7 pF

(5) Variable Capacitor Diode (SVC332 Sanyo)

(6) IFT (I) (CMFQ-021A Toko Co.,)



(7) IFT (II)



2150-208-033 Sumida Co., Cent. Freq. 455kHz Q≥95

7LC-4751B Toko Co., Cent Freq. 455kHz Q≥75

between 2 and 3 170t. Tuned Cap. 180pF between 2 and 3 146t. Tuned Cap. 180pF

(8) Narrow Band Resonator (BFB450C4 N Murata Co.,)

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